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## THE HARMONIC COMPOSITION OF CURRENT IN ZERO-WORKING WIRE WITH NON-LINEAR LOAD

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**ABSTRACT:** The purpose of this study was experimental investigations and a study of the nature of the change in the main harmonic components of the current in the neutral working wire of a three-phase four-wire network with a voltage of 0.38 kV. The study of the amplitude-phase-frequency characteristics of currents flowing in an electric network with a predominant non-linear load is one of the most urgent tasks, the solution of which will increase the efficiency of electric power transmission and eliminate the negative consequences due to the current flow in a zero-working wire in the network at 0.38 kV. To study the effect of load changes on the amplitude-phase-frequency characteristics of currents in the linear and zero working wires at the input of the load node, measurements were carried out by certified electrical measuring instruments namely, the "Resource-UF2M" and "Hioki 3196". With these devices, the phase voltages at the input to the load node were measured. The analysis of the results obtained for the load node whose power was formed mainly by a lighting system with fluorescent and LED lamps and a system of office electrical receivers (computers, copiers, printers, scanners, etc.) was performed. It can be concluded that a current comparable to the currents of the linear wires of the network flows from the load node with the predominant nonlinear power receivers through the zero-working wire. At the same time, in the zero-working wire of the network, the third harmonic current prevails over the main frequency currents.

**ABSTRAK:** Tujuan kajian ini adalah penyelidikan eksperimental dan kajian mengenai sifat perubahan komponen harmonik utama arus dalam wayar kerja neutral rangkaian empat wayar tiga fasa dengan voltan 0.38 kV. Kajian mengenai ciri-ciri frekuensi-amplitud-fasa arus yang mengalir dalam rangkaian elektrik dengan beban bukan linear yang dominan adalah salah satu tugas yang paling mendesak, penyelesaiannya akan meningkatkan kecekapan penghantaran tenaga elektrik dan menghilangkan akibat negatif yang disebabkan ke arus semasa dalam wayar sifar bekerja di rangkaian pada 0.38 kV. Untuk mengkaji kesan perubahan beban pada ciri frekuensi-fasa frekuensi arus dalam wayar linier dan sifar pada input nod beban, pengukuran dilakukan oleh alat pengukur elektrik yang disahkan iaitu, "Resource-UF2M" dan "Hioki 3196". Dengan peranti ini, voltan fasa pada input ke simpul beban diukur. Analisis hasil yang diperolehi untuk simpul beban yang kekuatannya dibentuk terutama oleh sistem pencahayaan dengan lampu pendarfluor dan LED dan sistem penerima elektrik pejabat (komputer, mesin fotokopi, pencetak, pengimbas, dll.). Dapat disimpulkan bahawa arus yang setanding dengan arus

wayar linier rangkaian mengalir dari nod beban dengan penerima kuasa bukan linier yang dominan melalui wayar sifar. Pada masa yang sama, dalam wayar sifar rangkaian, arus harmonik ketiga berlaku berbanding arus frekuensi utama.

**KEYWORDS:** *three-phase four-wire network; zero working wire of the network; harmonic currents; non-symmetry; non-linear power receiver*

## 1. INTRODUCTION

The main reasons for current appearance in the zero-working wire of a three-phase four-wire network are the asymmetry of the loads in the phases and presence of nonlinear power receivers. When flowing through a zero-working wire, the components of the currents of higher harmonics and asymmetries cause additional losses of voltage and electricity; reduction in the service life of cable lines; increase the resistance of grounding devices of electrical installations; and interference in low-voltage communication lines [1-6].

The study of the amplitude-phase-frequency characteristics of currents flowing in an electric network with a predominant non-linear load is one of the most urgent tasks, the solution of which will increase the efficiency of electric power transmission and eliminate the negative consequences due to the current flow in a zero-working wire in the network at 0.38 kV [7].

Since the spectral composition of currents in real electric networks is influenced by a large number of simultaneously influencing factors (the composition of power receivers in the load node, their connection diagram, switching frequency, etc.), the study of the harmonic composition of currents of the most common typical consumers is of practical interest. The power of the load node under study was formed mainly by a lighting system with fluorescent and LED lamps and a system of office electrical receivers (computers, copiers, printers, scanners, etc.) [8].

To study the effect of load changes on the amplitude-phase-frequency characteristics of currents in linear and zero working wires at the input of the load node, measurements were carried out by certified electrical measuring instruments [9]. The “Resource-UF2M” and “Hioki 3196” devices controlled currents in linear and zero working (“Hioki 3196”) wires. At the same time, the same devices measured the phase voltages at the input to the load node.

## 2. CHARACTERISTICS OF THE DISTORTION COEFFICIENT OF THE SINE CURVE IN ACCORDANCE WITH DAILY CHANGES

The change in the total active power at the input of the load node during the day is shown in Fig. 1. The range of changes in the total active power is large – from 5 kW at night up to 68 kW during the day. The greatest loads were during daylight hours from 7:30 to 17:00 hours. The minimum value of the total active power, as seen in Fig. 1, was recorded in the morning (from 5:00 to 7:00 hours), and the maximum – in the daytime from 12:00 to 13:32 hours. As can be seen from Fig. 1, the power consumption of the load node occurred according to a two-stage schedule.

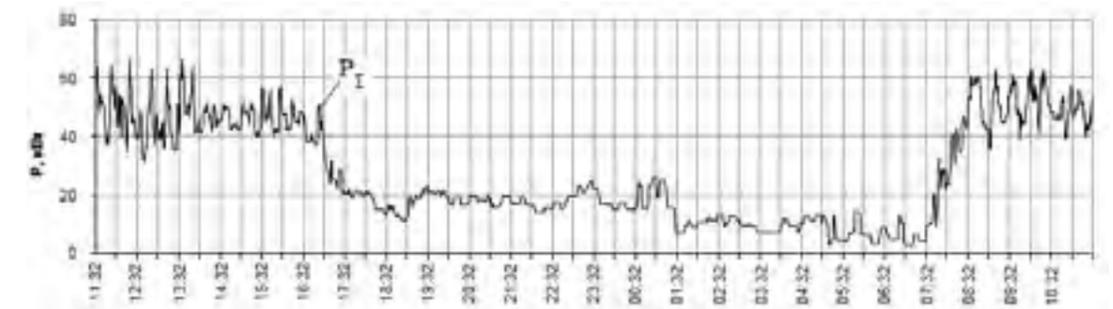


Fig. 1: Schedule of changes in total active power at the input of the load node.

Moreover, the change in active capacities during the day is very asymmetric in the network phases (Fig. 2). The active power  $P_1$  of the first phase L1 during the day turned out to be the smallest most of the time. Separate bursts of active power  $P_1$  of phase L1 were clearly observed in the period from 5:00 to 6:00 hours, when the load was the greatest. Most of the day, the greatest load fell on the L3 phase with an active power of  $P_3$ . Only in the period from 14:00 to 16:32 hours the active power of  $P_2$  exceeded  $P_3$ .

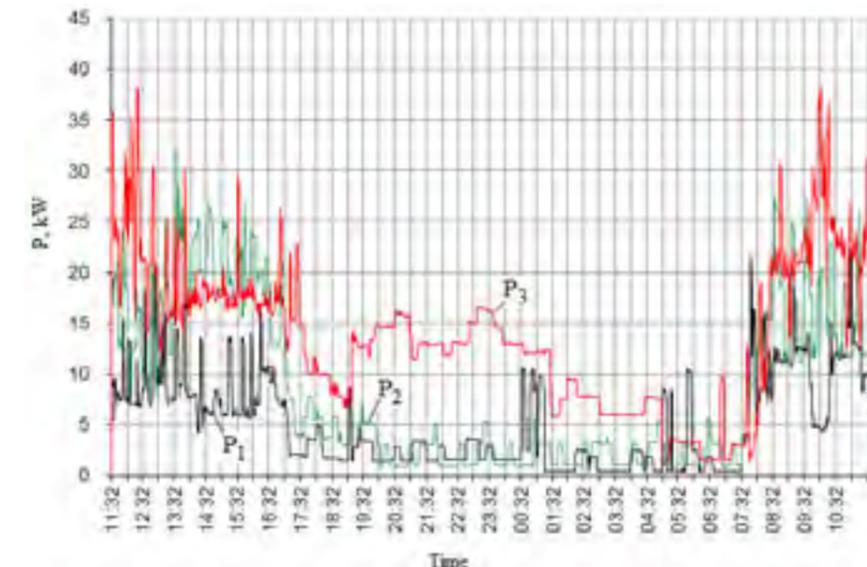


Fig. 2: Daily graphs of changes in active powers for the phases of the network 0.38 kV at the input of the load node.

Due to the asymmetric load of the phases, the voltage asymmetry coefficients in the zero and reverse sequences of the fundamental frequency were significant (Fig. 3). So, for example, the voltage asymmetry coefficient in the zero sequence of the fundamental frequency ( $K_{0U}$ ) did not fall during the day, even with a minimum (below 2%) load. During periods of greatest power consumption (from 8:00 to 17:00 hours), the voltage unbalance coefficient in the zero sequence of the fundamental frequency often exceeded 5%.

The voltage asymmetry coefficient in the reverse sequence of the fundamental frequency ( $K_{2U}$ ) during the day had lower values. Moreover, it turned out to be more stable than the coefficient  $K_{0U}$ . The range of variation of  $K_{2U}$  ranged from 2% to 4%. The results of these studies, as well as an analysis of literature [10-15] indicate that the active use of switching power supplies in electrical appliances leads to an increase in harmonic distortions of voltages and currents in the electrical network. Moreover, the distortion coefficients of

the sinusoidality of the voltages have lower values than the distortion coefficients of the sinusoidality of the load currents.

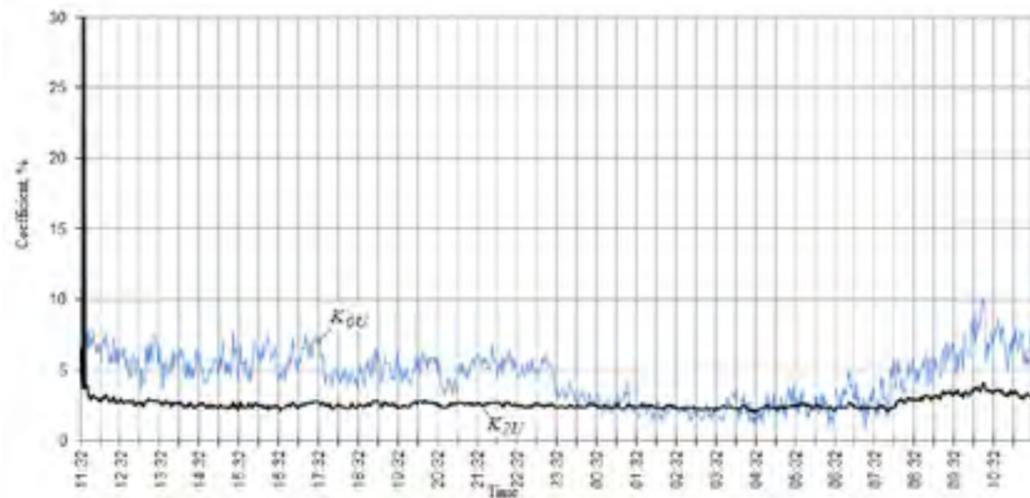


Fig. 3: Diagrams of daily changes in the voltage asymmetry coefficients in the zero and reverse sequence at the input of the load node with the prevailing non-linear power receivers.

Figure 4 shows graphs of daily changes in the distortion coefficients of the sinusoidality of the voltage curves in the phases of the network at the input of the load node 0.38 kV. It should be noted that in none of the phases at the input of the load node, the distortion coefficient of the sinusoidality of the voltage curves did not exceed the minimum allowable value – 8%.

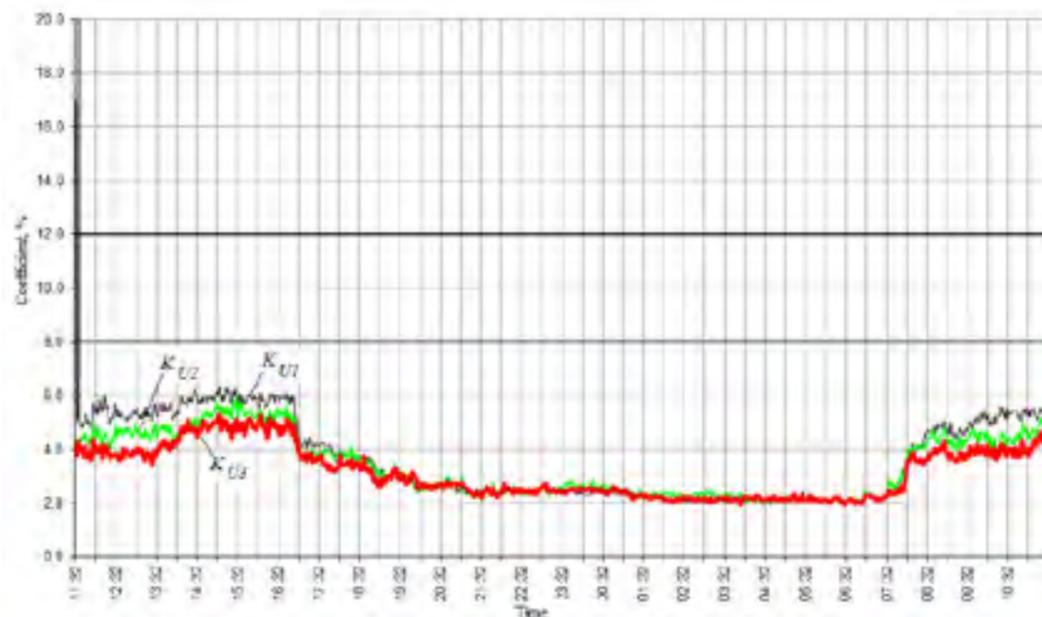


Fig. 4: Diagrams of daily changes in the distortion coefficients of the sinusoidality of the phase voltage curves at the input of the load node with predominant non-linear power receivers.

During periods of maximum load (from 8:00 to 17:00 hours), the distortion coefficient of the sinusoidality of the voltage curve in the first phase  $K_{U1}$  was the largest, more than 6% (see Fig. 4). During load shedding, the distortion coefficients of the sinusoidality of the voltage curves in phases became the same (over 2%). The average amplitude-frequency characteristic of the most significant average daily coefficients of the 3rd, 5th, 7th, 9th and 11th voltage harmonics in the phases of the network 0.38 kV at the input to the load node (Fig. 5) clearly show that the odd harmonic voltage components prevail over even ones.

Among the odd higher voltage harmonics, the 3rd, 5th, and 9th harmonics had the largest amplitudes. It should be noted that the largest amplitudes of the 3rd and 5th voltage harmonics were recorded in the first phase, and the 9th voltage harmonic was the largest in the second phase of the network 0.38 kV (Fig. 5). The coefficients of the 3rd voltage harmonic by phase at the input to the load node varied from 4.7% to 5.5%, the coefficients of the 5th voltage harmonic, respectively, from 1.4% to 2.5%, and the coefficients of the 9th voltage harmonic, respectively, from 1.0% to 1.4% (Fig. 5).

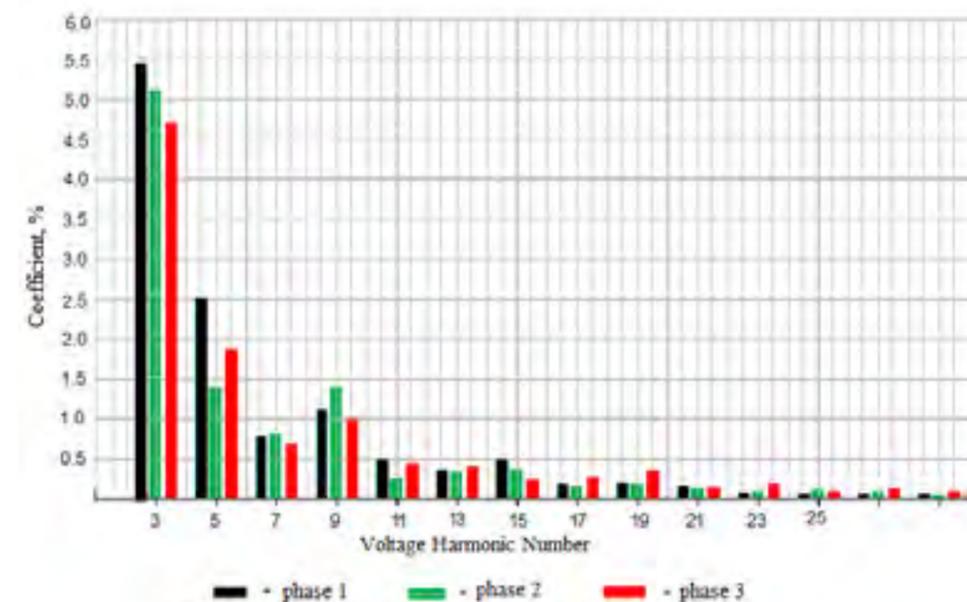


Fig. 5: Frequency response of the most significant daily average coefficients of the  $n$ -th harmonic components of the phase voltage at the input of the load node.

### 3. CHANGES IN CURRENT PARAMETERS AT VARIOUS LOADS OF THE POWER RECEIVER

To assess the degree of distortion of currents sinusoidality during the period of the highest loads, the waveforms of the phase currents and in the zero-working wire at the input of the load node were recorded using the “Hioki 3196” instrument (Fig. 6).

The current waveforms correspond to the time interval on the daily graph of the change in the total active power at the highest load (from 11:40 to 12:00 hours according to the schedule of Fig. 1). An analysis of the current shape in the zero-working wire  $i_4(t)$  on the waveform shows (Fig. 6) that it is not substantially sinusoidal. The waveforms of currents in the linear wires of the load node  $i_1(t)$ ,  $i_2(t)$  and  $i_3(t)$  are also significantly distorted by higher harmonics.

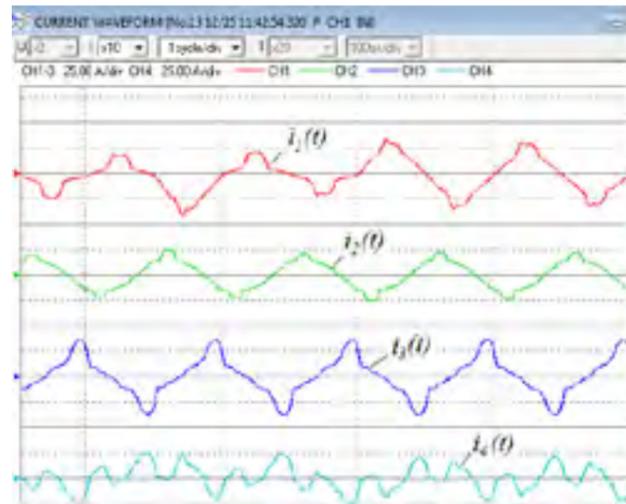


Fig. 6: A fragment of the waveforms of the phase currents and the zero-working wire at the input of the load node with the prevailing non-linear power receivers.

Using the “Hioki 3196” software package, the amplitudes and phases of the currents were determined for seven significant harmonic components not equal to zero (odd harmonics from the 1st to the 13th). For compact recordings, these transformations are presented in the form of Fig. 7, which also shows mathematical models (expressions) of instantaneous current values at the input to the load node.

In Fig. 7, the Fourier expansion is performed up to the 50th harmonic of the current. This allows us to justify the selection of the seven most significant harmonic components of the current by comparison. The sinusoidal distortion coefficients of the curves (see Fig. 7) are denoted by THD (Total Harmonic Distortions) and are called the “total current harmonic distortion coefficient”.

As can be seen from Fig. 5, the values of the sinusoidality distortion coefficients of the current curves THDI1, THDI2 and THDI3 significantly exceed the similar voltage coefficients. In the first linear wire L1 at the input of the load node, the total THDI current distortion factor for the considered time was 23.7%, in the second wire L2 – 15.98%, in the third wire L3 – 36.18%, and in the neutral wire – 138.35%, respectively.

The analysis of the harmonic composition of the current of the zero-working wire  $i_4(t)$  is of interest. As can be seen from Fig. 1, in the working (mean square) current  $I_4$ , the proportion of the main harmonic of the current is 6.77 A. The proportion of the third harmonic components of the currents flowing in the linear wires of the network 0.38 kV, in the zero working wire for the considered time point, is 8.9 A, in 5th harmonic current, respectively, 2.33 A, 7th – 0.79 A, 9th – 1.13 A, 11th – 0.51 A and 13th – 0.58 A. In this case, the total current load of the zero-working wire for the considered time was equal to 12.48 A.

With this load mode, the current values in the linear wires of the network 0.38 kV were as follows:  $I_1 = 16.63$  A;  $I_2 = 13.54$  A;  $I_3 = 18.8$  A. That is, the current load of the zero working wire is commensurate with the load of the linear wires. Obviously, the third and ninth current harmonics make the main contribution to the increase in the current load of the zero-working wire to the values of the currents flowing in the linear wires of the network 0.38 kV.

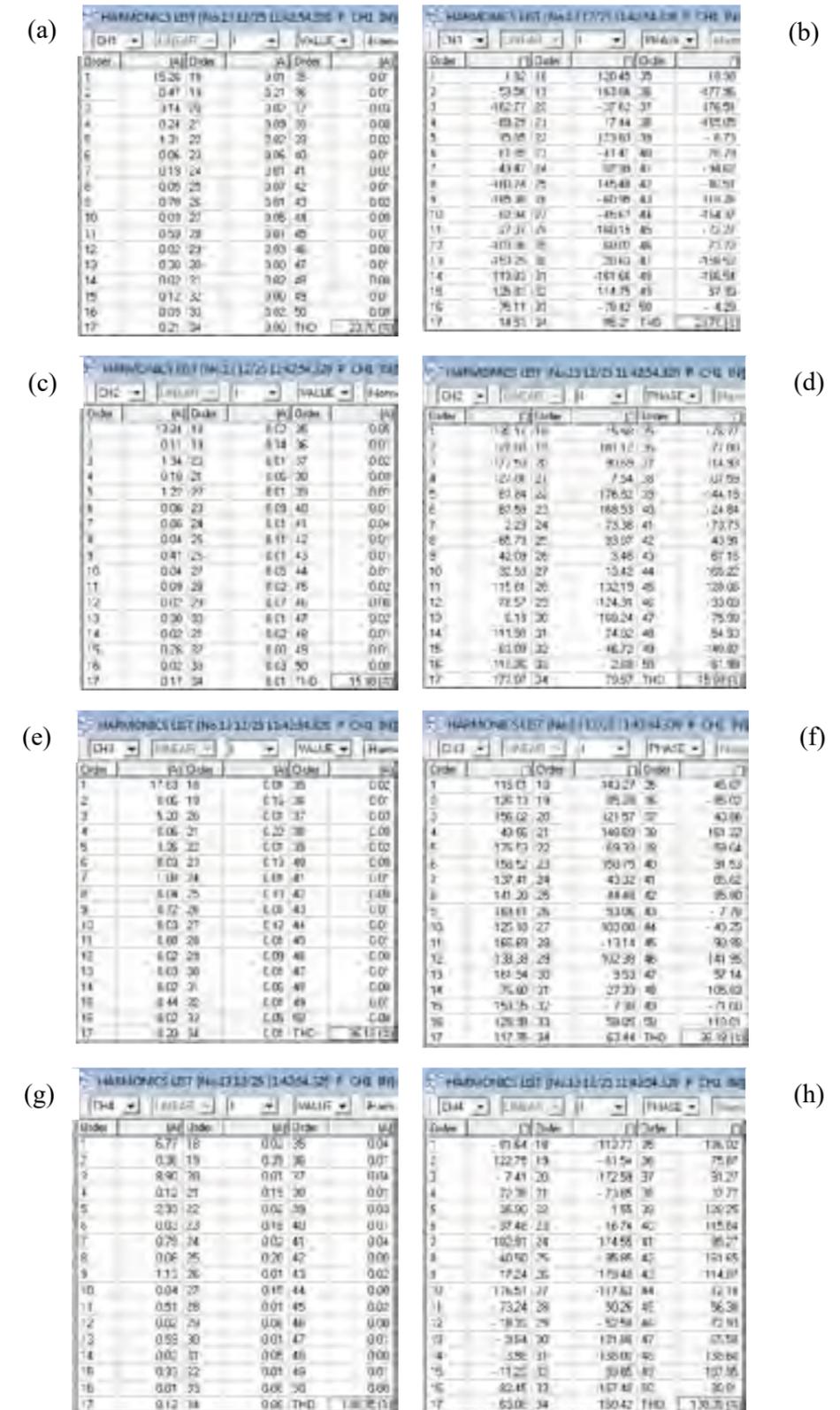


Fig. 7: Vector diagrams of currents and voltages at a frequency of 150 Hz at the input of a load node with predominant nonlinear power consumers:

$$\begin{aligned}
 \text{a), b) } - i_1(t) &= \sqrt{2} \cdot \left( \begin{aligned} &15,26 \sin(\omega t + 2^\circ) + 3,14 \sin(3\omega t - 163^\circ) + \\ &+1,31 \sin(5\omega t + 16^\circ) + 0,19 \sin(7\omega t - 44^\circ) + 0,78 \sin(9\omega t - 165^\circ) + \\ &+0,59 \sin(11\omega t + 27^\circ) + 0,30 \sin(13\omega t - 153^\circ) \end{aligned} \right) \text{ A} \\
 \text{c), d) } - i_2(t) &= \sqrt{2} \cdot \left( \begin{aligned} &13,34 \sin(\omega t - 130^\circ) + 1,34 \sin(3\omega t - 173^\circ) + \\ &+1,22 \sin(5\omega t + 88^\circ) + 0,86 \sin(7\omega t + 2^\circ) + 0,47 \sin(9\omega t - 42^\circ) + \\ &+0,09 \sin(11\omega t + 116^\circ) + 0,30 \sin(13\omega t - 0^\circ) \end{aligned} \right) \text{ A} \\
 \text{e), f) } - i_3(t) &= \sqrt{2} \cdot \left( \begin{aligned} &17,63 \sin(\omega t + 115^\circ) + 5,20 \sin(3\omega t + 156^\circ) + \\ &+3,26 \sin(5\omega t - 177^\circ) + 1,08 \sin(7\omega t - 137^\circ) + 0,72 \sin(9\omega t + 164^\circ) + \\ &+0,68 \sin(11\omega t + 167^\circ) + 0,69 \sin(13\omega t + 162^\circ) \end{aligned} \right) \text{ A} \\
 \text{g), h) } - i_4(t) &= \sqrt{2} \cdot \left( \begin{aligned} &6,77 \sin(\omega t - 82^\circ) + 8,90 \sin(3\omega t - 7^\circ) + \\ &+2,33 \sin(5\omega t - 37^\circ) + 0,79 \sin(7\omega t + 103^\circ) + 1,13 \sin(9\omega t + 17^\circ) + \\ &+0,51 \sin(11\omega t - 73^\circ) + 0,59 \sin(13\omega t - 10^\circ) \end{aligned} \right) \text{ A}
 \end{aligned}$$

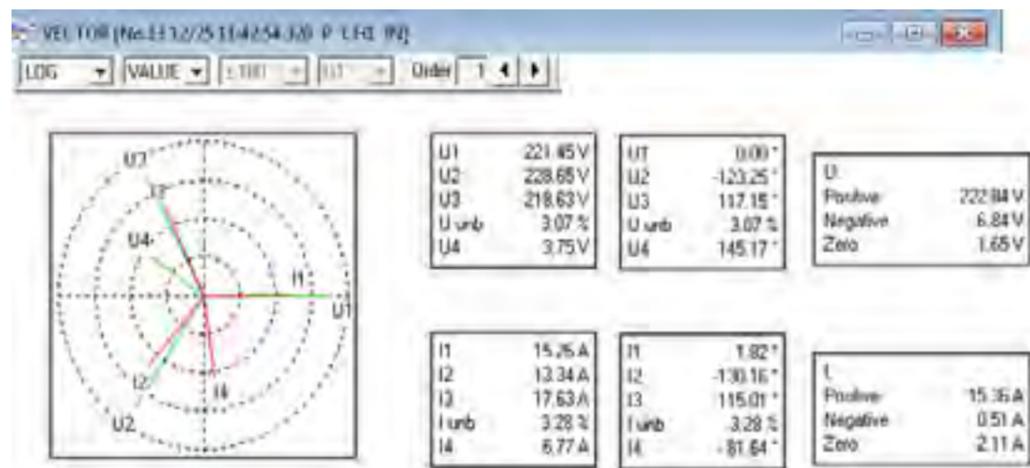


Fig. 8: Vector diagrams of currents and voltages at a frequency of 50 Hz at the input of a load node with predominant nonlinear power consumers.

Figure 8, obtained using the “HIOKI 3196” instrument software, shows the logical vector diagrams of currents and voltages at the input of the load node with prevailing nonlinear power receivers at time 11:42 (see Fig. 1). Along with this, the amplitude of the third harmonic of the current  $i_4(t)$  is greater than the amplitude of the fundamental harmonic.

In addition to the current values of voltage and current, the phase shifts of each value are also indicated here. For the first harmonic (frequency 50 Hz) the components of the direct (Positive), reverse (Negative) and zero (Zero) sequences of voltages and currents are given. With these components, real three-phase voltage and current systems can be balanced to a symmetrical system. Similar characteristics for the considered time and frequency of 150 Hz are presented in Fig. 9.

A distinctive feature of the vector diagrams of voltages and currents of the third harmonic (Fig. 9) is the lack of common mode between the vectors of different phases. Along with the lack of phase matching of vectors, they are also characterized by inequality in magnitude. So, for example, the effective current of the third harmonic in the first phase is 3.14 A, in the second phase is 1.34 A, and in the third phase is 5.2 A (see Fig. 1). This is

explained by the nature of the formation of these currents in the load node in case of accidental switching on of various power consumers under different voltages.

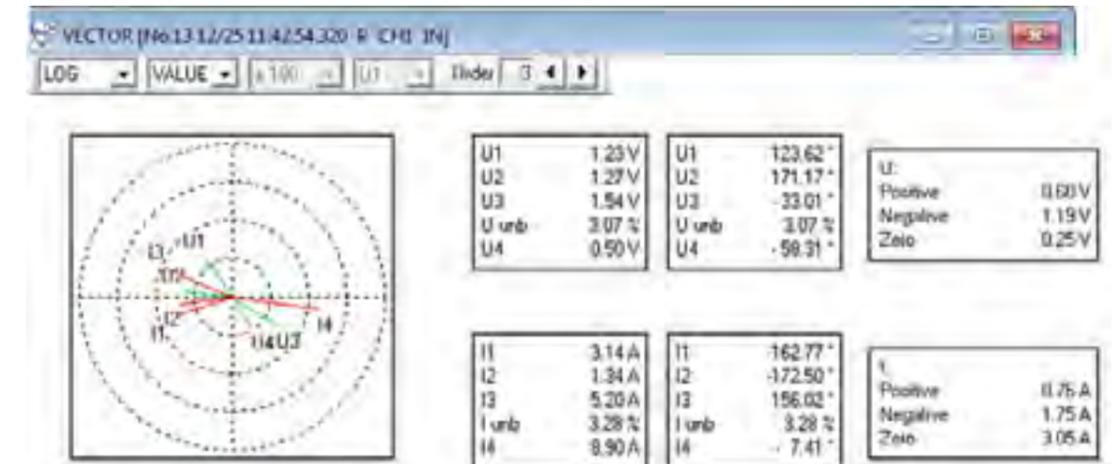


Fig. 9: Vector diagrams of currents and voltages at a frequency of 150 Hz at the input of a load node with predominant nonlinear power consumers.

#### 4. CONCLUSIONS

The research contribution of this work lies in the fact that an experimental study was carried out, the nature of the change in the fundamental harmonic components of the current in the zero-working wire of a three-phase four-wire network with a voltage of 0.38 kV was studied. The results obtained during the work will increase the efficiency of electric power transmission and eliminate the negative consequences due to the current flow in a zero-working wire in the network 0.38 kV.

The analysis of the research results shows that in the presence of non-linear power receivers, the current in the zero-working wire of a three-phase four-wire network is comparable with the currents in linear wires. So, in the maximum load mode, with the current value of the zero working wire current 12.48 A, the currents of the linear wires were  $I_1 = 16.63$  A,  $I_2 = 13.54$  A, and  $I_3 = 18.8$  A, respectively.

Based on the obtained results of the study, it can be argued that the currents of the third harmonic component, which dominate in the neutral working wire of the network 0.38 kV, are not in-phase in the linear wires. The results obtained are of great practical importance when designing technical means that provide current unloading of the zero-working wire of the network 0.38 kV and, ultimately, reduce energy losses during transmission.

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## CHARACTERIZATION AND MELT SPINNING OF POLY (LACTIC ACID)/POLY (ETHYLENE GLYCOL) BLENDS

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**ABSTRACT:** In this study, melt blended compositions of pure PLA with additions of polyethylene glycol (PEG) up to 30 wt% were prepared. Fourier-transform infrared spectroscopy (FTIR), differential scanning calorimeter (DSC), and thermogravimetric analysis (TGA) were used to investigate the properties of PLA/PEG blends, such as structural, thermal, and morphological properties. The results showed that further increments of PEG cause the -OH group of PLA/PEG blends to show a broad peak, indicating that there is hydrogen bonding interaction between PEG and PLA chains. DSC result revealed that the addition of PEG decreases the glass transition temperature from 57 °C to 46 °C and crystallization temperature from 107 °C to 87 °C. Such trends suggest enhanced chain mobility of PLA chains. TGA thermograms showed that further additions of PEG into PLA resulted in a consistent shift to lower temperature and decrease in thermal stability. Optical microscopy (OM) and scanning electron microscopy (SEM) observations of the melt spun PLA/PEG microfibers revealed that the diameter of the microfibers averaged between 15 to 80 microns.

**ABSTRAK:** Kajian ini menganalisa komposisi adunan lebur PLA asli bersama tambahan polietilena glikol (PEG) sebanyak 30%. Penjelmaan Fourier spektroskopi inframerah (FTIR), kalorimeter pengimbasan pembezaan (DSC) dan analisis termogravimetri (TGA) telah digunakan bagi mengkaji sifat-sifat adunan PLA/PEG, seperti struktur, terma dan sifat-sifat morfologi. Keputusan menunjukkan penambahan PEG seterusnya menyebabkan kumpulan -OH campuran PLA/PEG memberikan puncak yang lebar, ini menunjukkan ada interaksi ikatan hidrogen antara rangkaian PEG dan PLA. Keputusan DSC menunjukkan penambahan PEG mengurangkan perubahan gelas dari 57 °C kepada 46 °C dan suhu kristalisasi dari 107 °C kepada 87 °C. Trend ini mencadangkan peningkatan pergerakan rangkaian pada rangkaian PLA. Termogram TGA menunjukkan dengan penambahan berterusan PEG ke dalam PLA menghasilkan penurunan konsisten pada suhu dan pengurangan kestabilan haba. Pemerhatian mikroskop optik (OM) dan mikroskopi elektron pensakanan (SEM) mikrofiber spun lebur PLA/PEG menunjukkan purata diameter mikrofiber ini antara 15 ke 80 mikron.

**KEYWORDS:** *microfiber; poly lactic acid (PLA); polyethylene glycol (PEG); melt spinning*